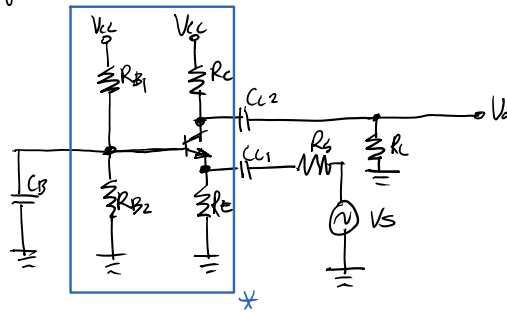


Common Base Amplifier

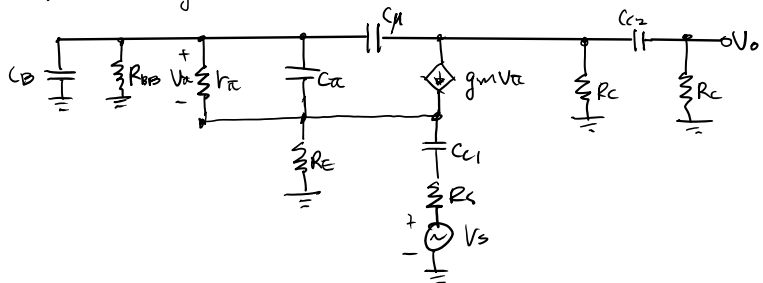
October 17, 2017 3:48 PM

Typical common-base amplifier:

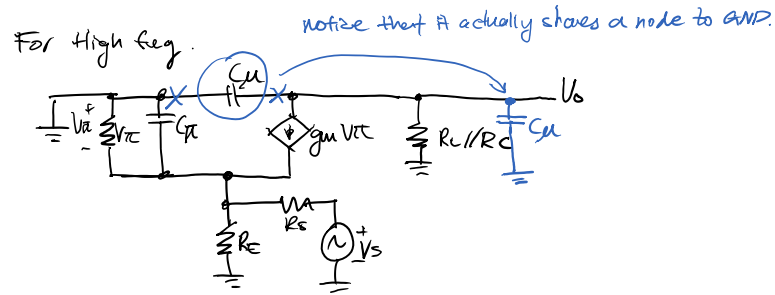
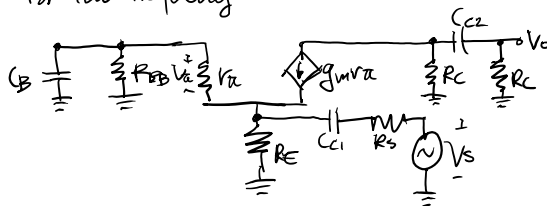


Assuming that β is biased using $\frac{1}{2}$ Rule, (Same way for Common Emitter Amplifiers)
 $\rightarrow R_{B1}, R_{B2}, R_C, R_E$ is known.

The Small Signal model is:



For low frequency



Midband:

$$V_{be} = -\frac{(\frac{1}{1+\beta})r_{\pi} \parallel R_E}{R_s + [(\frac{1}{1+\beta})r_{\pi} \parallel R_E]} V_s$$

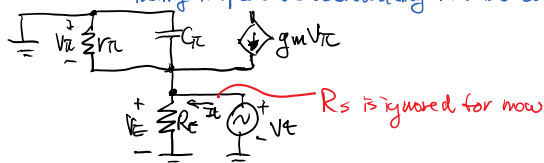
$$\approx -\frac{(\frac{1}{1+\beta})r_{\pi}}{R_s + \frac{r_{\pi}}{1+\beta}} V_s$$

$$A_m = \frac{V_o}{V_s} = g_m(R_C \parallel R_L) \cdot V_{be}$$

* Notice that common base amplifiers are non-inverting

High Frequency:

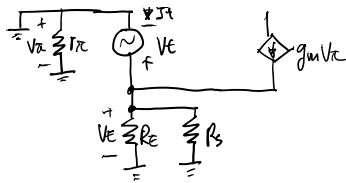
Finding impedance seen looking into the emitter



$$Z_E = (R_E \parallel \frac{r_{\pi}}{1+\beta} \parallel \frac{1}{sC_E})$$

notice that C_E is unaffected





$$I_t = \frac{V_t}{R_E} + \frac{V_t}{R_s} - \frac{V_{be}}{R_C} - g_m V_{be}, \text{ and } V_{be} = -V_t$$

$$= V_t \left(\frac{1}{R_E} + \frac{1}{R_s} + \frac{1}{R_C} + g_m \right), \text{ and } g_m = \frac{\beta}{r_{\pi}}$$

$$= V_t \left(R_E \parallel R_s \parallel \frac{r_{\pi}}{1+\beta} \right)$$

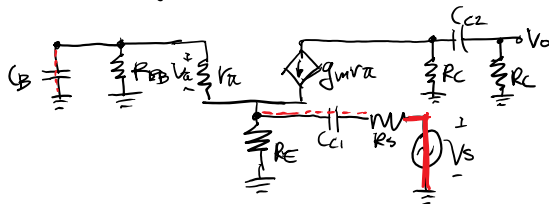
$$\text{Thus } \frac{I_t}{V_t} = Y_E = R_E \parallel R_s \parallel \frac{r_{\pi}}{1+\beta}$$

Low Frequency

Immediately, we can recognize C_C and it's pole is:

$$\omega_{p1} = \frac{1}{C_C(R_C + R_L)}$$

For the remaining poles, we need to do SCTC tests.



For the collector side, no SCTC test is required.

For the base side, do SCTC and OCTC tests:

$$\begin{cases} \tau_{sc}^{CB} = C_B \cdot (R_{BB} \parallel (r_{\pi} + (1+\beta) R_E \parallel R_s)) \\ \tau_{sc}^{CC1} = C_{C1} (R_s + R_E \parallel \frac{r_{\pi}}{1+\beta}) \leftarrow \text{More resistance} \end{cases}$$

$$\begin{cases} \tau_{oc}^{CB} = C_B \cdot (R_{BB} \parallel [r_{\pi} + (1+\beta) R_E]) \\ \tau_{oc}^{CC1} = C_{C1} \cdot \left(\frac{r_{\pi} + R_{BB}}{1+\beta} \parallel R_E + R_s \right) \end{cases}$$